University of Illinois at Urbana-Champaign
Dept. of Electrical and Computer Engineering

## ECE 120: Introduction to Computing

Boolean Properties and Optimization

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## The Dual Form Swaps 0/1 and AND/OR

Boolean algebra has an interesting property called duality.
Let's define the dual form of an expression as follows:

- Starting with the expression,
- swap 0 with 1
(just the values, not variables),
- and swap AND with OR.

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## Every Boolean Expression Has a Dual Form

For example, what is the dual of

$$
\mathrm{A}+(\mathrm{BC})+(0(\mathrm{D}+1)) ?
$$

First replace the 0 with 1 and the 1 with 0 .
Then replace $+(\mathrm{OR})$ with (AND)
and vice-versa.
We obtain:

$$
A \cdot(B+C) \cdot(1+(D \cdot 0))
$$

## The Dual of the Dual is the Expression

So what is the dual of

$$
A \cdot(B+C) \cdot(1+(D \cdot 0)) ?
$$

Since we're swapping things, swapping them again produces the original expression:

$$
A+(B C)+(0(D+1))
$$

Thus any Boolean expression has a unique dual, and the dual of the dual is the expression (hence the term duality-two aspects of the same thing).

## Pitfall: Do Not Change the Order of Operations

Be careful not to change the order of operations when finding a dual form.
For example, the dual form of

$$
\mathrm{A}+\mathrm{BC}
$$

is

$$
\mathrm{A}(\mathrm{~B}+\mathrm{C})
$$

The operation on $\mathbf{B}$ and $\mathbf{C}$ must happen before the other operation.

## Why Do You Care? One Reason: the Principle of Duality

Three reasons:

- CMOS gate structures are dual forms
- Quick way to complement any expression
- the principle of duality

Let's start with the last, which we'll use shortly (when we examine more properties).
Principle of duality: If a Boolean theorem or identity is true/false, so is the dual of that theorem or identity.
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## Generalized DeMorgan is Quick and Easy

Let's say that we have an expression F .
To find F' ... apply DeMorgan's Laws ...
Apply repeatedly, as many times as necessary.
Or use the generalized version based on duality:

- Write the dual form of $\mathbf{F}$.
- Swap variables and
complemented variables.
- (That's all.)


## An Example of Finding a Complement with the Dual Form

$$
\begin{gathered}
\mathrm{F}=\mathrm{AB}\left(\mathrm{C}+\mathrm{DL}^{\prime} \mathrm{G}\left(\mathrm{~B}^{\prime}+\mathrm{A}+\mathrm{E}\right)\right)\left(\mathrm{H}+\mathrm{J}^{\prime} \mathrm{A}^{\prime} \mathrm{B}\right) \\
\text { What's } \mathrm{F}^{\prime} ?
\end{gathered}
$$

The dual is

$$
\mathrm{A}+\mathrm{B}+\underset{\left.\left(\mathrm{H} \mathrm{~J}^{\prime}+\mathrm{A}^{\prime}+\mathrm{B}\right)^{\left(\mathrm{B}^{\prime} \mathrm{AE}\right)}\right)+}{\left(\mathrm{C} \mathrm{~L}^{\prime}+\mathrm{G}+\left({ }^{(2)}\right.\right.}
$$

So

$$
\left.\left.\mathrm{F}^{\prime}=\mathbf{A}^{\prime}+\mathrm{B}^{\prime}+\underset{\left(\mathrm{H}^{\prime}\right.}{ }\left(\mathrm{H}^{\prime}\left(\mathrm{D}^{\prime}+\mathrm{A}+\mathrm{L}^{\prime}+\mathrm{B}^{\prime}\right)\right)\left(\mathbf{B A}^{\prime} \mathrm{E}^{\prime}\right)\right)\right)+
$$

You can skip the middle step once you're comfortable with the process.

## We Can Derive a Gate's Output from the n-type Network

What about CMOS gate structures?
Think about the network of n-type
MOSFETS connecting an output Q to 0 V .
For example, consider a set of
four n-type arranged in parallel with inputs A, B, C, and D.
So $\mathbf{Q}=0$ if ANY of the transistors is on. In other words, $\mathbf{Q}$ is 0 when $\mathbf{A}+\mathbf{B}+\mathbf{C}+\mathrm{D}$.
Thus $\mathbf{Q}=(\mathbf{A}+\mathbf{B}+\mathbf{C}+\mathbf{D})^{\prime}$. A NOR gate.

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## We Can Also Derive Function from the p-type Network

What about the p-type transistors on the same gate?
${ }^{\circ}$ They are arranged in series.
-They connect $\mathbf{Q}$ to $\mathrm{V}_{\mathrm{dd}}$.
But p-type transistors are on when their gates are set to 0 . So $\mathbf{Q}=1$ when ALL of the inputs are 0 .
Thus $\mathbf{Q}=\mathbf{A}^{\prime} \mathbf{B}^{\prime} \mathbf{C}^{\prime} \mathbf{D}^{\prime}$.
That's the same expression, of course.

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## The Expressions are Related via Generalized DeMorgan

But notice that we can also

- get the second form
- by applying generalized DeMorgan to the first form.
Starting with

$$
\mathbf{Q}=(\mathbf{A}+\mathrm{B}+\mathrm{C}+\mathrm{D})^{\prime},
$$

we find the dual of $A+B+C+D$ to be $A B C D$, so

$$
\mathbf{Q}=A^{\prime} \mathbf{B}^{\prime} \mathbf{C}^{\prime} \mathbf{D}^{\prime} .
$$

## The Networks are Dual Forms of One Another

The complemented variables come from the use of p-type transistors.
The dual form is built into the gate design.
If we want to design a gate for something OTHER than NAND, NOR, NOT:

- Write the output as $\mathbf{Q}=(\text { expression })^{\prime}$,
- Build that expression from n-type MOSFETs.
- Build the dual of the expression from p-type MOSFETs.



## An Example of an Unusual Gate

Consider the gate here:
From the n-type network,

$$
\mathbf{Q}=((\mathbf{A}+\mathrm{B}) \mathbf{C})^{\prime}
$$

The dual of the expression (ignoring the complement) is

$$
\mathrm{AB}+\mathrm{C}
$$

which is the structure of the p-type network.


## Optimization versus Abstraction

Most designers just use NAND and NOR
(or, today, even higher-level abstractions!).
In general:

- breaking abstraction boundaries
can give us an advantage,
- but the boundaries make
the design task less complex,
- which improves human productivity and reduces the likelihood of mistakes.
That's another tradeoff.
Computer aided design (CAD) tools can perform some of these optimizations for us, too.


## Area and Speed for the Unusual Gate

So the function $\mathbf{Q}=((\mathbf{A}+\mathbf{B}) \mathbf{C})^{\prime}$ requires six transistors and one gate delay.
We can, of course, limit ourselves
to NAND/NOR gates.
In that case, $\mathbf{Q}=\left(\left(\mathrm{A}^{\prime} \mathrm{B}^{\prime}\right)^{\prime} \mathbf{C}\right)^{\prime}$
We use one two-input NAND for ( $\left.A^{\prime} B^{\prime}\right)^{\prime}$, and a second two-input NAND for $\mathbf{Q}$.
If we assume that $A^{\prime}$ and $B^{\prime}$ are available, the NAND design requires eight transistors and two gate delays.

## More Dual Form Boolean Properties

DeMorgan's Laws are also dual forms

$$
(A+B)^{\prime}=A^{\prime} B^{\prime} \quad(A B)^{\prime}=A^{\prime}+B^{\prime}
$$

What about distributivity? Here's the rule that you know from our usual algebra

$$
A(B+C)=A B+A C
$$

(multiplication distributes over addition)
It's also true in Boolean algebra:
AND distributes over OR.

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## OR Also Distributes Over AND in Boolean Algebra

$$
A(B+C)=A B+A C
$$

Now take the dual form...

$$
\mathrm{A}+\mathrm{BC}=(\mathrm{A}+\mathrm{B})(\mathrm{A}+\mathrm{C})
$$

OR distributes over AND!
(Note that this property does NOT hold in our usual algebra. $14+7 \cdot 4 \neq(14+7)(14+4))$
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## One More Property: Consensus

The last property is non-intuitive.

$$
\mathrm{AB}+\mathrm{A}^{\prime} \mathrm{C}+\mathrm{BC}=\mathrm{AB}+\mathrm{A}^{\prime} \mathrm{C}
$$

It's called "consensus" because

- the first two terms TOGETHER (when both are true, and thus reach a consensus) imply the third term
- so the third term can be dropped.


## A K-Map Illustrates Consensus Well

Let's look at a K-map.
AB is the vertical green loop.
$\mathrm{A}^{\prime} \mathrm{C}$ is the horizontal green loop.
$\mathbf{B C}$ is the black loop.


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## Consensus Has Two Dual Forms (SOP and POS)

And, of course, there is another form of consensus for POS form.

Start with our first form:

$$
\mathrm{AB}+\mathrm{A}^{\prime} \mathrm{C}+\mathrm{BC}=\mathrm{AB}+\mathrm{A}^{\prime} \mathrm{C}
$$

Then find the dual to obtain:
$(A+B)\left(A^{\prime}+C\right)(B+C)=(A+B)\left(A^{\prime}+C\right)$

